

MCP73831/2

Miniature Single-Cell, Fully Integrated Li-Ion, Li-Polymer Charge Management Controllers

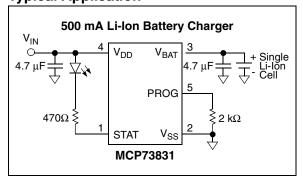
Features:

- Linear Charge Management Controller:
 - Integrated Pass Transistor
 - Integrated Current Sense
 - Reverse Discharge Protection
- High Accuracy Preset Voltage Regulation: <u>+</u> 0.75%
- Four Voltage Regulation Options:
 - 4.20V, 4.35V, 4.40V, 4.50V
- Programmable Charge Current
- Selectable Preconditioning
- Selectable End-of-Charge Control
- · Charge Status Output
 - Tri-State Output MCP73831
 - Open-Drain Output MCP73832
- · Automatic Power-Down
- Thermal Regulation
- Temperature Range: -40°C to +85°C
- · Packaging:
 - 8-Lead. 2 mm x 3 mm DFN
 - 5-Lead, SOT23

Applications:

- · Lithium-Ion/Lithium-Polymer Battery Chargers
- Personal Data Assistants
- · Cellular Telephones
- Digital Cameras
- MP3 Players
- · Bluetooth Headsets
- · USB Chargers

Typical Application



Description:

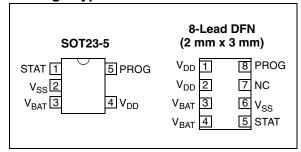
The MCP73831/2 devices are highly advanced linear charge management controllers for use in space-limited, cost-sensitive applications. The MCP73831/2 are available in an 8-Lead, 2 mm x 3 mm DFN package or a 5-Lead, SOT23 package. Along with their small physical size, the low number of external components required make the MCP73831/2 ideally suited for portable applications. For applications charging from a USB port, the MCP73831/2 adhere to all the specifications governing the USB power bus.

The MCP73831/2 employ a constant-current/constant-voltage charge algorithm with selectable preconditioning and charge termination. The constant voltage regulation is fixed with four available options: 4.20V, 4.35V, 4.40V or 4.50V, to accommodate new, emerging battery charging requirements. The constant current value is set with one external resistor. The MCP73831/2 devices limit the charge current based on die temperature during high power or high ambient conditions. This thermal regulation optimizes the charge cycle time while maintaining device reliability.

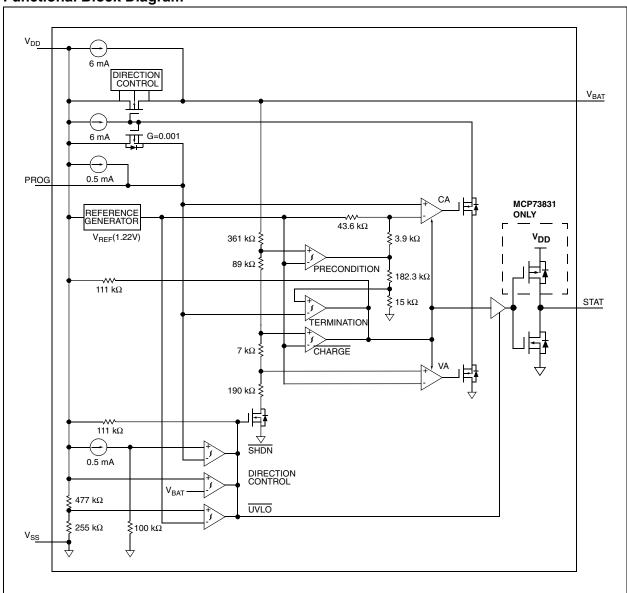
Several options are available for the preconditioning threshold, preconditioning current value, charge termination value and automatic recharge threshold. The preconditioning value and charge termination value are set as a ratio, or percentage, of the programmed constant current value. Preconditioning can be disabled. Refer to Section 1.0 "Electrical Characteristics" for available options and the "Product Identification System" for standard options.

The MCP73831/2 devices are fully specified over the ambient temperature range of -40°C to +85°C.

Package Types



Functional Block Diagram



1.0 **ELECTRICAL CHARACTERISTICS**

Absolute Maximum Ratings†

V _{DD} 7.0V
All Inputs and Outputs w.r.t. $V_{SS} \ldots \ldots$ -0.3 to $(V_{DD} \! + \! 0.3)V$
$\label{eq:maximum Junction Temperature} \ \ T_J Internally \ Limited$
Storage temperature65°C to +150°C
ESD protection on all pins:
Human Body Model (1.5 k $\!\Omega$ in Series with 100 pF) $\!\ge\!4$ kV
Machine Model (200 pF, No Series Resistance)400V

† Notice: Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

DC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG}(typ.) + 0.3V]$ to 6V, $T_A = -40$ °C to +85°C. Typical values are at +25°C, $V_{DD} = [V_{REG} (typ.) + 1.0V]$ **Parameters** Sym. Max. Units Conditions Min. Тур. Supply Input Supply Voltage V_{DD} 3.75 6 ٧ Supply Current 1500 510 μΑ Charging I_{SS} 53 200 μΑ Charge Complete. No Battery **PROG Floating** 25 50 μΑ $V_{DD} \leq (V_{BAT} - 50 \text{ mV})$ 5 μΑ 1 2 0.1 $V_{DD} < V_{STOP}$ μΑ V_{DD} Low-to-High **UVLO Start Threshold** 3.45 3.6 ٧ V_{START} 3.3 **UVLO Stop Threshold** 3.38 ٧ V_{DD} High-to-Low V_{STOP} 3.2 3.5 **UVLO** Hysteresis 70 m۷ V_{HYS} **Voltage Regulation (Constant-Voltage Mode)** Regulated Output Voltage 4.168 4.20 4.232 MCP7383X-2 V_{REG} 4.317 4.35 4.383 ٧ MCP7383X-3 4.367 4.40 4.433 ٧ MCP7383X-4 MCP7383X-5 4.466 4.50 4.534 V $V_{DD} = [V_{REG}(Typ) + 1V]$ $I_{OUT} = 10 \text{ mA}$ $T_A = -5^{\circ}\text{C to } +55^{\circ}\text{C}$ %/V Line Regulation $|(\Delta V_{BAT}/V_{BAT}$ 0.09 $V_{DD} = [V_{REG}(Typ)+1V]$ to 6V 0.30 ΔV_{DD} $I_{OUT} = 10 \text{ mA}$ $I_{OUT} = 10 \text{ mA to } 50 \text{ mA}$ 0.05 Load Regulation $|\Delta V_{BAT}/V_{BAT}|$ 0.30 $V_{DD} = [V_{REG}(Typ)+1V]$ Supply Ripple Attenuation **PSRR** 52 dΒ I_{OUT}=10 mA, 10Hz to 1 kHz 47 dΒ I_{OUT} =10 mA, 10Hz to 10 kHz 22 dΒ I_{OUT} =10 mA, 10Hz to 1 MHz **Current Regulation (Fast Charge Constant-Current Mode)** Fast Charge Current 100 110 PROG = 10 $k\Omega$ 90 mΑ I_{REG} Regulation 505 450 550 PROG = $2.0 \text{ k}\Omega$, Note 1 mΑ $T_A = -5^{\circ}C$ to $+55^{\circ}C$

Not production tested. Ensured by design. Note 1:

DC CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG}(typ.) + 0.3V]$ to 6V, $T_A = -40^{\circ}C$ to $+85^{\circ}C$. Typical values are at $+25^{\circ}C$, $V_{DD} = [V_{REG}(typ.) + 1.0V]$

Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions
Preconditioning Current	Regulation (Tri	ckle Charge Co	nstant-Current	Mode)		
Precondition Current	I _{PREG} / I _{REG}	7.5	10	12.5	%	PROG = 2.0 kΩ to 10 kΩ
Ratio		15	20	25	%	PROG = 2.0 kΩ to 10 kΩ
		30	40	50	%	PROG = 2.0 kΩ to 10 kΩ
		_	100	_	%	No Preconditioning
						$T_A = -5^{\circ}C$ to $+55^{\circ}C$
Precondition Voltage	V _{PTH} / V _{REG}	64	66.5	69	%	V _{BAT} Low-to-High
Threshold Ratio		69	71.5	74	%	V _{BAT} Low-to-High
Precondition Hysteresis	V _{PHYS}	_	110	_	mV	V _{BAT} High-to-Low
Charge Termination					•	
Charge Termination	I _{TERM} / I _{REG}	3.75	5	6.25	%	PROG = 2.0 kΩ to 10 kΩ
Current Ratio		5.6	7.5	9.4	%	PROG = 2.0 kΩ to 10 kΩ
		7.5	10	12.5	%	PROG = 2.0 kΩ to 10 kΩ
		15	20	25	%	PROG = 2.0 kΩ to 10 kΩ
						$T_A = -5^{\circ}C \text{ to } +55^{\circ}C$
Automatic Recharge					•	
Recharge Voltage	V _{RTH} / V _{REG}	91.5	94.0	96.5	%	V _{BAT} High-to-Low
Threshold Ratio		94	96.5	99	%	V _{BAT} High-to-Low
Pass Transistor ON-Resi	stance			•	•	
ON-Resistance	R _{DSON}	_	350	_	mΩ	$V_{DD} = 3.75V, T_J = 105^{\circ}C$
Battery Discharge Curre					•	
Output Reverse Leakage	I _{DISCHARGE}	_	0.15	2	μА	PROG Floating
Current		_	0.25	2	μΑ	V _{DD} Floating
		_	0.15	2	μΑ	V _{DD} < V _{STOP}
		_	-5.5	-15	μΑ	Charge Complete
Status Indicator – STAT					•	
Sink Current	I _{SINK}	_	_	25	mA	
Low Output Voltage	V _{OL}	_	0.4	1	V	I _{SINK} = 4 mA
Source Current	I _{SOURCE}	_	_	35	mA	
High Output Voltage	V _{OH}	_	V _{DD} -0.4	V _{DD} - 1	V	I _{SOURCE} = 4 mA (MCP73831)
Input Leakage Current	I _{LK}	_	0.03	1	μΑ	High-Impedance
PROG Input					•	
Charge Impedance Range	R _{PROG}	2	_	20	kΩ	
Minimum Shutdown Impedance	R _{PROG}	70	_	200	kΩ	
Automatic Power Down	ı				1	1
Automatic Power Down Entry Threshold	V _{PDENTER}	V _{DD} <(V _{BAT} +20mV)	V _{DD} <(V _{BAT} +50mV)	_		$3.5V \le V_{BAT} \le V_{REG}$ V_{DD} Falling
Automatic Power Down Exit Threshold	V _{PDEXIT}	_	V _{DD} <(V _{BAT} +150mV)	V _{DD} <(V _{BAT} +200mV)		$3.5V \le V_{BAT} \le V_{REG}$ $V_{DD} \text{ Rising}$
Thermal Shutdown	ı				1	1
Die Temperature	T _{SD}	_	150	_	°C	
Die Temperature Hysteresis	T _{SDHYS}	_	10	_	°C	

Note 1: Not production tested. Ensured by design.

AC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG} \text{ (typ.)} + 0.3V]$ to 12V, $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$. Typical values are at $+25^{\circ}\text{C}$, $V_{DD} = [V_{REG} \text{ (typ.)} + 1.0V]$									
Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions			
UVLO Start Delay	t _{START}	_		5	ms	V _{DD} Low-to-High			
Constant-Current Regulation	n								
Transition Time Out of Preconditioning	t _{DELAY}	_	_	1	ms	$V_{BAT} < V_{PTH}$ to $V_{BAT} > V_{PTH}$			
Current Rise Time Out of Preconditioning	t _{RISE}	_	_	1	ms	I _{OUT} Rising to 90% of I _{REG}			
Termination Comparator Filter	t _{TERM}	0.4	1.3	3.2	ms	Average I _{OUT} Falling			
Charge Comparator Filter	t _{CHARGE}	0.4	1.3	3.2	ms	Average V _{BAT}			
Status Indicator									
Status Output turn-off	t _{OFF}	_	_	200	μS	I _{SINK} = 1 mA to 0 mA			
Status Output turn-on	t _{ON}	_		200	μS	I _{SINK} = 0 mA to 1 mA			

TEMPERATURE SPECIFICATIONS

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG} \text{ (typ.)} + 0.3V]$ to 12V. Typical values are at +25°C, $V_{DD} = [V_{REG} \text{ (typ.)} + 1.0V]$									
Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions			
Temperature Ranges									
Specified Temperature Range	T _A	-40		+85	°C				
Operating Temperature Range	TJ	-40	_	+125	°C				
Storage Temperature Range	T _A	-65	_	+150	°C				
Thermal Package Resistances									
5-Lead, SOT23	θ_{JA}	_	230	_	°C/W	4-Layer JC51-7 Standard Board, Natural Convection			
8-Lead, 2 mm x 3 mm, DFN	θ_{JA}	_	76	_	°C/W	4-Layer JC51-7 Standard			

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

NOTE: Unless otherwise indicated, $V_{DD} = [V_{REG}(typ.) + 1V]$, $I_{OUT} = 10$ mA and $T_A = +25$ °C, Constant-Voltage mode.

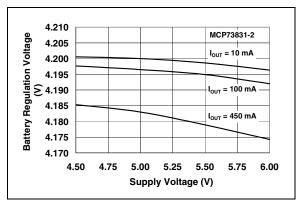


FIGURE 2-1: Battery Regulation Voltage (V_{BAT}) vs. Supply Voltage (V_{DD}) .

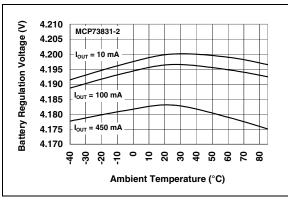


FIGURE 2-2: Battery Regulation Voltage (V_{BAT}) vs. Ambient Temperature (T_A) .

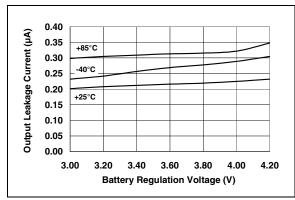


FIGURE 2-3: Output Leakage Current $(I_{DISCHARGE})$ vs. Battery Regulation Voltage (V_{BAT}) .

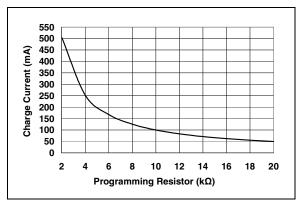


FIGURE 2-4: Charge Current (I_{OUT}) vs. Programming Resistor (R_{PROG}).

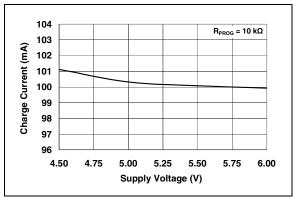


FIGURE 2-5: Charge Current (I_{OUT}) vs. Supply Voltage (V_{DD}) .

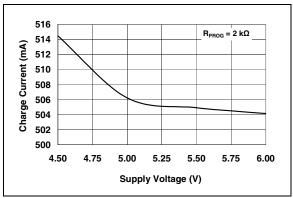


FIGURE 2-6: Charge Current (I_{OUT}) vs. Supply Voltage (V_{DD}) .

TYPICAL PERFORMANCE CURVES (CONTINUED)

NOTE: Unless otherwise indicated, $V_{DD} = [V_{REG}(typ.) + 1V]$, $I_{OUT} = 10$ mA and $T_{A} = +25$ °C, Constant-Voltage mode.

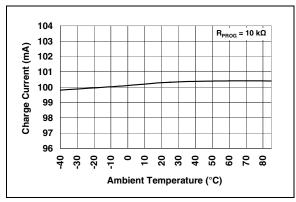


FIGURE 2-7: Charge Current (I_{OUT}) vs. Ambient Temperature (T_A) .

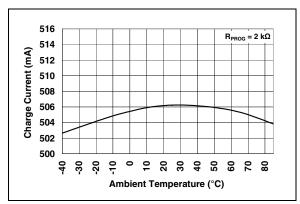


FIGURE 2-8: Charge Current (I_{OUT}) vs. Ambient Temperature (T_A) .

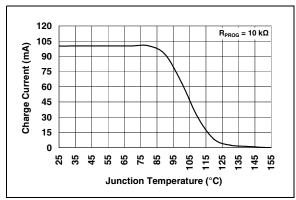


FIGURE 2-9: Charge Current (I_{OUT}) vs. Junction Temperature (T_{ij}) .

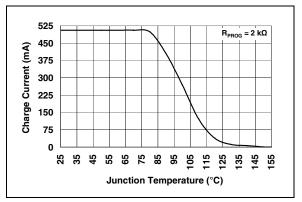


FIGURE 2-10: Charge Current (I_{OUT}) vs. Junction Temperature (T_J) .

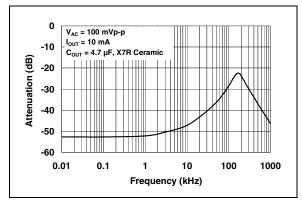


FIGURE 2-11: Power Supply Ripple Rejection (PSRR).

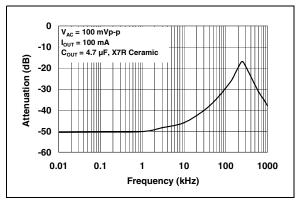


FIGURE 2-12: Power Supply Ripple Rejection (PSRR).

TYPICAL PERFORMANCE CURVES (CONTINUED)

NOTE: Unless otherwise indicated, $V_{DD} = [V_{REG}(typ.) + 1V]$, $I_{OUT} = 10$ mA and $T_A = +25$ °C, Constant-Voltage mode.

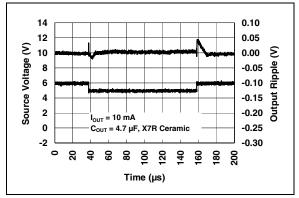


FIGURE 2-13: Line Transient Response.

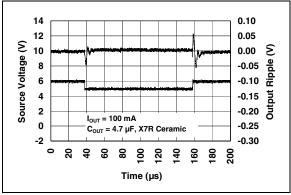


FIGURE 2-14: Line Transient Response.

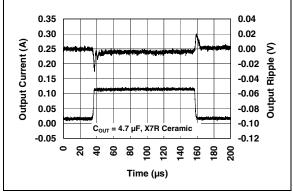


FIGURE 2-15: Load Transient Response.

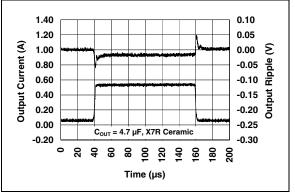


FIGURE 2-16: Load Transient Response.

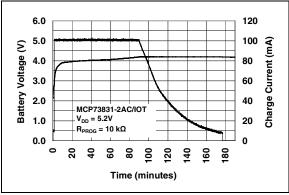


FIGURE 2-17: Complete Charge Cycle (180 mAh Li-Ion Battery).

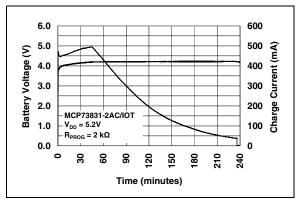


FIGURE 2-18: Complete Charge Cycle (1000 mAh Li-Ion Battery).

3.0 PIN DESCRIPTION

The descriptions of the pins are listed in Table 3-1.

TABLE 3-1: PIN FUNCTION TABLES

Pin No.		Counch al	Formation					
DFN	SOT23-5	Symbol	Function					
1	4	V_{DD}	Battery Management Input Supply					
2	_	V_{DD}	Battery Management Input Supply					
3	3	V_{BAT}	Battery Charge Control Output					
4	_	V _{BAT}	Battery Charge Control Output					
5	1	STAT	Charge Status Output					
6	2	V _{SS}	Battery Management 0V Reference					
7	_	NC	No Connection					
8	5	PROG	Current Regulation Set and Charge Control Enable					

3.1 Battery Management Input Supply (V_{DD})

A supply voltage of [V_{REG} (typ.) + 0.3V] to 6V is recommended. Bypass to V_{SS} with a minimum of 4.7 $\mu F.$

3.2 Battery Charge Control Output (V_{BAT})

Connect to positive terminal of battery. Drain terminal of internal P-channel MOSFET pass transistor. Bypass to V_{SS} with a minimum of 4.7 μF to ensure loop stability when the battery is disconnected.

3.3 Charge Status Output (STAT)

STAT is an output for connection to an LED for charge status indication. Alternatively, a pull-up resistor can be applied for interfacing to a host microcontroller.

STAT is a tri-state logic output on the MCP73831 and an open-drain output on the MCP73832.

3.4 Battery Management 0V Reference (V_{SS})

Connect to negative terminal of battery and input supply.

3.5 Current Regulation Set (PROG)

Preconditioning, fast charge and termination currents are scaled by placing a resistor from PROG to V_{SS} .

The charge management controller can be disabled by allowing the PROG input to float.



4.0 DEVICE OVERVIEW

The MCP73831/2 are highly advanced linear charge management controllers. Figure 4-1 depicts the operational flow algorithm from charge initiation to completion and automatic recharge.

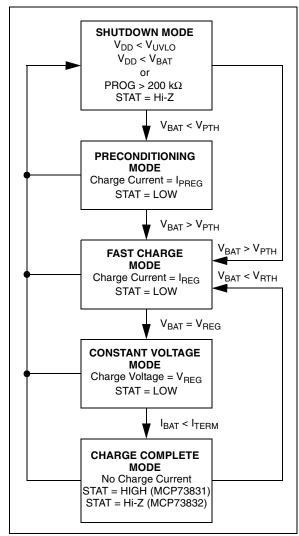


FIGURE 4-1: Flowchart.

4.1 Under Voltage Lockout (UVLO)

An internal UVLO circuit monitors the input voltage and keeps the charger in Shutdown mode until the input supply rises above the UVLO threshold. The UVLO circuitry has a built in hysteresis of 100 mV.

In the event a battery is present when the input power is applied, the input supply must rise 150 mV above the battery voltage before the MCP73831/2 becomes operational.

The UVLO circuit places the device in Shutdown mode if the input supply falls to within +50 mV of the battery voltage. Again, the input supply must rise to a level 150 mV above the battery voltage before the MCP73831/2 become operational.

The UVLO circuit is always active. At any time the input supply is below the UVLO threshold or within +50~mV of the voltage at the V_{BAT} pin, the MCP73831/2 are placed in a Shutdown mode.

During any UVLO condition, the battery reverse discharge current shall be less than 2 μ A.

4.2 Charge Qualification

For a charge cycle to begin, all UVLO conditions must be met and a battery or output load must be present. A charge current programming resistor must be connected from PROG to V_{SS} . If the PROG pin is open or floating, the MCP73831/2 are disabled and the battery reverse discharge current is less than 2 μ A. In this manner, the PROG pin acts as a charge enable and can be used as a manual shutdown.

4.3 Preconditioning

If the voltage at the V_{BAT} pin is less than the preconditioning threshold, the MCP73831/2 enter a preconditioning or Trickle Charge mode. The preconditioning threshold is factory set. Refer to **Section 1.0** "**Electrical Characteristics**" for preconditioning threshold options and the Product Identification System for standard options.

In this mode, the MCP73831/2 supply a percentage of the charge current (established with the value of the resistor connected to the PROG pin) to the battery. The percentage or ratio of the current is factory set. Refer to **Section 1.0** "**Electrical Characteristics**" for preconditioning current options and the Product Identification System for standard options.

When the voltage at the V_{BAT} pin rises above the preconditioning threshold, the MCP73831/2 enter the Constant-Current or Fast Charge mode.

4.4 Fast Charge Constant-Current Mode

During the Constant-Current mode, the programmed charge current is supplied to the battery or load. The charge current is established using a single resistor from PROG to V_{SS} . Constant-Current mode is maintained until the voltage at the V_{BAT} pin reaches the regulation voltage, V_{REG} .



4.5 Constant-Voltage Mode

When the voltage at the V_{BAT} pin reaches the regulation voltage, V_{REG} , constant voltage regulation begins. The regulation voltage is factory set to 4.2V, 4.35V, 4.40V, or 4.50V with a tolerance of $\pm 0.75\%$.

4.6 Charge Termination

The charge cycle is terminated when, during Constant-Voltage mode, the average charge current diminishes below a percentage of the programmed charge current (established with the value of the resistor connected to the PROG pin). A 1 ms filter time on the termination comparator ensures that transient load conditions do not result in premature charge cycle termination. The percentage or ratio of the current is factory set. Refer to **Section 1.0 "Electrical Characteristics"** for charge termination current options and the "**Product Identification System**" for standard options.

The charge current is latched off and the MCP73831/2 enter a Charge Complete mode.

4.7 Automatic Recharge

The MCP73831/2 continuously monitor the voltage at the V_{BAT} pin in the Charge Complete mode. If the voltage drops below the recharge threshold, another charge cycle begins and current is once again supplied to the battery or load. The recharge threshold is factory set. Refer to **Section 1.0** "Electrical Characteristics" for recharge threshold options and the Product Identification System for standard options.

4.8 Thermal Regulation

The MCP73831/2 limit the charge current based on the die temperature. The thermal regulation optimizes the charge cycle time while maintaining device reliability. Figure 4-2 depicts the thermal regulation for the MCP73831/2.

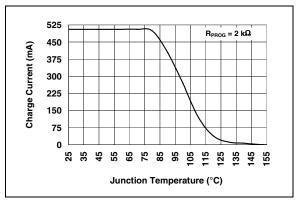


FIGURE 4-2: Thermal Regulation.

4.9 Thermal Shutdown

The MCP73831/2 suspend charge if the die temperature exceeds 150°C. Charging will resume when the die temperature has cooled by approximately 10°C.

5.0 DETAILED DESCRIPTION

5.1 Analog Circuitry

5.1.1 BATTERY MANAGEMENT INPUT SUPPLY (V_{DD})

The V_{DD} input is the input supply to the MCP73831/2. The MCP73831/2 automatically enter a Power-Down mode if the voltage on the V_{DD} input falls below the UVLO voltage (V_{STOP}). This feature prevents draining the battery pack when the V_{DD} supply is not present.

5.1.2 CURRENT REGULATION SET (PROG)

Fast charge current regulation can be scaled by placing a programming resistor (R_{PROG}) from the PROG input to V_{SS} . The program resistor and the charge current are calculated using the following equation:

$$I_{REG} = \frac{1000V}{R_{PROG}}$$

Where:

 R_{PROG} = kOhms I_{REG} = milliampere

The preconditioning trickle charge current and the charge termination current are ratiometric to the fast charge current based on the selected device options.

5.1.3 BATTERY CHARGE CONTROL OUTPUT (V_{BAT})

The battery charge control output is the drain terminal of an internal P-channel MOSFET. The MCP73831/2 provide constant current and voltage regulation to the battery pack by controlling this MOSFET in the linear region. The battery charge control output should be connected to the positive terminal of the battery pack.

5.2 Digital Circuitry

5.2.1 STATUS INDICATOR (STAT)

The charge status output of the MCP73831 has three different states: High (H), Low (L), and High-Impedance (Hi-Z). The charge status output of the MCP73832 is open-drain, and, as such, has two different states: Low (L), and High-Impedance (Hi-Z). The charge charge status output can be used to illuminate 1, 2, or tri-color LEDs. Optionally, the charge status output can be used as an interface to a host microcontroller.

Table 5-1 summarize the state of the status output during a charge cycle..

TABLE 5-1: STATUS OUTPUT

Charge Cycle State	STAT1			
Charge Cycle State	MCP73831	MCP73832		
Shutdown	Hi-Z	Hi-Z		
No Battery Present	Hi-Z	Hi-Z		
Preconditioning	L	L		
Constant-Current Fast Charge	L	L		
Constant Voltage	L	L		
Charge Complete – Standby	Н	Hi-Z		

5.2.2 DEVICE DISABLE (PROG)

The current regulation set input pin (PROG) can be used to terminate a charge at any time during the charge cycle, as well as to initiate a charge cycle or initiate a recharge cycle.

Placing a programming resistor from the PROG input to V_{SS} enables the device. Allowing the PROG input to float or by applying a logic-high input signal, disables the device and terminates a charge cycle. When disabled, the device's supply current is reduced to 25 μ A, typically.

6.0 APPLICATIONS

The MCP73831/2 are designed to operate in conjunction with a host microcontroller or in a stand-alone application. The MCP73831/2 provide the preferred charge algorithm for Lithium-lon and Lithium-Polymer

cells constant current followed by constant voltage. Figure 6-1 depicts a typical stand-alone application circuit, while Figures 6-2 and 6-3 depict the accompanying charge profile.

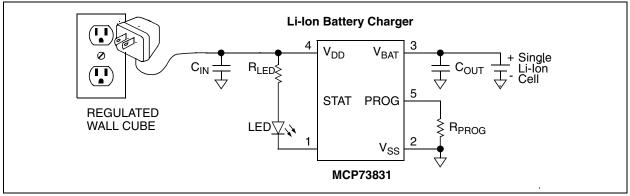


FIGURE 6-1: Typical Application Circuit.

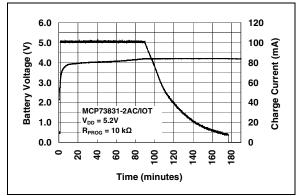


FIGURE 6-2: Typical Charge Profile (180 mAh Battery).

6.0 600 5.0 500 Battery Voltage (V) 400 4.0 3.0 300 2.0 200 ge MCP73831-2AC/IO 100 1.0 $V_{nn} = 5.2V$ $R_{PROG} = 2 k\Omega$ 0.0 210 8 8 8 8 50 8 Time (minutes)

FIGURE 6-3: Typical Charge Profile in Thermal Regulation (1000 mAh Battery).

6.1 Application Circuit Design

Due to the low efficiency of linear charging, the most important factors are thermal design and cost, which are a direct function of the input voltage, output current and thermal impedance between the battery charger and the ambient cooling air. The worst-case situation is when the device has transitioned from the Preconditioning mode to the Constant-Current mode. In this situation, the battery charger has to dissipate the maximum power. A trade-off must be made between the charge current, cost and thermal requirements of the charger.

6.1.1 COMPONENT SELECTION

Selection of the external components in Figure 6-1 is crucial to the integrity and reliability of the charging system. The following discussion is intended as a guide for the component selection process.

6.1.1.1 Current Programming Resistor (R_{PROG})

The preferred fast charge current for Lithium-lon cells is at the 1C rate, with an absolute maximum current at the 2C rate. For example, a 500 mAh battery pack has a preferred fast charge current of 500 mA. Charging at this rate provides the shortest charge cycle times without degradation to the battery pack performance or life.

6.1.1.2 Thermal Considerations

The worst-case power dissipation in the battery charger occurs when the input voltage is at the maximum and the device has transitioned from the Preconditioning mode to the Constant-Current mode. In this case, the power dissipation is:

$$PowerDissipation = (V_{DDMAX} - V_{PTHMIN}) \times I_{REGMAX}$$

Where:

V_{DDMAX} = the maximum input voltage

 I_{REGMAX} = the maximum fast charge current

V_{PTHMIN} = the minimum transition threshold

voltage

Power dissipation with a 5V, $\pm 10\%$ input voltage source is:

$$PowerDissipation = (5.5V - 2.7V) \times 550mA = 1.54W$$

This power dissipation with the battery charger in the SOT23-5 package will cause thermal regulation to be entered as depicted in Figure 6-3. Alternatively, the 2mm x 3mm DFN package could be utilized to reduce charge cycle times.

6.1.1.3 External Capacitors

The MCP73831/2 are stable with or without a battery load. In order to maintain good AC stability in the Constant-Voltage mode, a minimum capacitance of 4.7 μF is recommended to bypass the V_{BAT} pin to $V_{SS}.$ This capacitance provides compensation when there is no battery load. In addition, the battery and interconnections appear inductive at high frequencies. These elements are in the control feedback loop during Constant-Voltage mode. Therefore, the bypass capacitance may be necessary to compensate for the inductive nature of the battery pack.

Virtually any good quality output filter capacitor can be used, independent of the capacitor's minimum Effective Series Resistance (ESR) value. The actual value of the capacitor (and its associated ESR) depends on the output load current. A 4.7 μF ceramic, tantalum or aluminum electrolytic capacitor at the output is usually sufficient to ensure stability for output currents up to a 500 mA.

6.1.1.4 Reverse-Blocking Protection

The MCP73831/2 provide protection from a faulted or shorted input. Without the protection, a faulted or shorted input would discharge the battery pack through the body diode of the internal pass transistor.

6.1.1.5 Charge Inhibit

The current regulation set input pin (PROG) can be used to terminate a charge at any time during the charge cycle, as well as to initiate a charge cycle or initiate a recharge cycle.

Placing a programming resistor from the PROG input to V_{SS} enables the device. Allowing the PROG input to float or by applying a logic-high input signal, disables the device and terminates a charge cycle. When disabled, the device's supply current is reduced to 25 μ A, typically.

6.1.1.6 Charge Status Interface

A status output provides information on the state of charge. The output can be used to illuminate external LEDs or interface to a host microcontroller. Refer to Table 5-1 for a summary of the state of the status output during a charge cycle.

6.2 PCB Layout Issues

For optimum voltage regulation, place the battery pack as close as possible to the device's V_{BAT} and V_{SS} pins. This is recommended to minimize voltage drops along the high current-carrying PCB traces.

If the PCB layout is used as a heatsink, adding many vias in the heatsink pad can help conduct more heat to the backplane of the PCB, thus reducing the maximum junction temperature. Figures 6-4 and 6-5 depict a typical layout with PCB heatsinking.

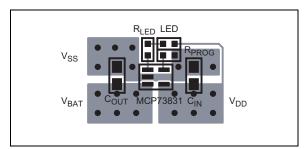


FIGURE 6-4: Typical Layout (Top).

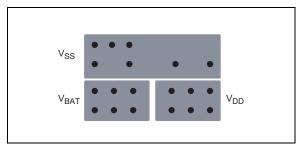


FIGURE 6-5: Typical Layout (Bottom).

7.0 PACKAGING INFORMATION

7.1 Package Marking Information

8-Lead DFN (2 mm x 3 mm)



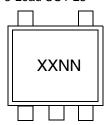
Device	Code
MCP73831T-2ACI/MC	AAE
MCP73831T-2ATI/MC	AAF
MCP73831T-2DCI/MC	AAG
MCP73831T-3ACI/MC	AAH
MCP73831T-4ADI/MC	AAJ
MCP73831T-5ACI/MC	AAK
MCP73832T-2ACI/MC	AAL
MCP73832T-2ATI/MC	AAM
MCP73832T-2DCI/MC	AAP
MCP73832T-3ACI/MC	AAQ
MCP73832T-4ADI/MC	AAR
MCP73832T-5ACI/MC	AAS
Mate. Applies to C. Lood	DEN

Note: Applies to 8-Lead DFN





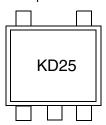
5-Lead SOT-23



Device	Code
MCP73831T-2ACI/OT	KDNN
MCP73831T-2ATI/OT	KENN
MCP73831T-2DCI/OT	KFNN
MCP73831T-3ACI/OT	KGNN
MCP73831T-4ADI/OT	KHNN
MCP73831T-5ACI/OT	KJNN
MCP73832T-2ACI/OT	KKNN
MCP73832T-2ATI/OT	KLNN
MCP73832T-2DCI/OT	KMNN
MCP73832T-3ACI/OT	KPNN
MCP73832T-4ADI/OT	KQNN
MCP73832T-5ACI/OT	KRNN

Note: Applies to 5-Lead SOT-23

Example:



Legend: XX...X Customer-specific information

Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')
NNN Alphanumeric traceability code

e3 Pb-free JEDEC designator for Matte Tin (Sn)

This package is Pb-free. The Pb-free JEDEC designator (e3)

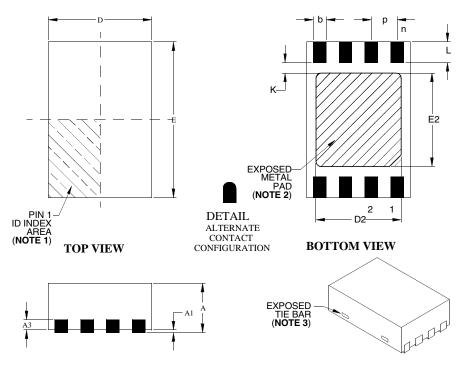
can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available

characters for customer-specific information.



8-Lead Plastic Dual-Flat, No-Lead Package (MC) 2x3x0.9 mm Body (DFN) - Saw Singulated



	Units		INCHES		MILLIMETERS*			
Dimension Limi	MIN	NOM	MAX	MIN	NOM	MAX		
Number of Pins	n		8			8		
Pitch	е		.020 BSC		0.50 BSC			
Overall Height	Α	.031	.035	.039	0.80	0.90	1.00	
Standoff	A1	.000	.001	.002	0.00	0.02	0.05	
Contact Thickness	А3		.008 REF.		0.20 REF.			
Overall Length	D		.079 BSC			2.00 BSC		
Overall Width	E		.118 BSC			3.00 BSC		
Exposed Pad Length	D2	.051	_	.069	1.30**	_	1.75	
Exposed Pad Width	E2	.059	_	.075	1.50**	_	1.90	
Contact Length §	L	.012 .016 .020			0.30	0.40	0.50	
Contact-to-Exposed Pad §	K	.008	_	-	0.20	_	-	
Contact Width	b	.008	.010	.012	0.20	0.25	0.30	

^{*} Controlling Parameter

§ Significant Characteristic

Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Exposed pad may vary according to die attach paddle size.
- 3. Package may have one or more exposed tie bars at ends.

BSC: Basic Dimension. Theoretically exact value shown without tolerances. See ASME Y14.5M

REF: Reference Dimension, usually without tolerance, for information purposes only. See ASME Y14.5M

JEDEC Equivalent MO-229 VCED-2

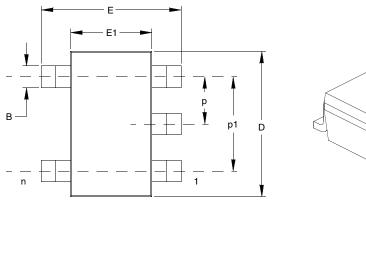
DWG No. C04-123

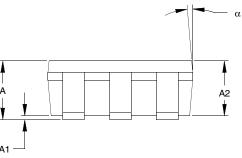
Revised 09-12-05

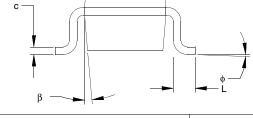


^{**} Not within JEDEC parameters

5-Lead Plastic Small Outline Transistor (OT) (SOT-23)







		INCHES*		N			
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		5			5	
Pitch	р		.038			0.95	
Outside lead pitch (basic)	p1		.075			1.90	
Overall Height	Α	.035	.046	.057	0.90	1.18	1.45
Molded Package Thickness	A2	.035	.043	.051	0.90	1.10	1.30
Standoff	A1	.000	.003	.006	0.00	0.08	0.15
Overall Width	E	.102	.110	.118	2.60	2.80	3.00
Molded Package Width	E1	.059	.064	.069	1.50	1.63	1.75
Overall Length	D	.110	.116	.122	2.80	2.95	3.10
Foot Length	L	.014	.018	.022	0.35	0.45	0.55
Foot Angle	f	0	5	10	0	5	10
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.014	.017	.020	0.35	0.43	0.50
Mold Draft Angle Top	а	0	5	10	0	5	10
Mold Draft Angle Bottom	b	0	5	10	0	5	10

^{*} Controlling Parameter

Notes

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. EIAJ Equivalent: SC-74A

Drawing No. C04-091

Revised 09-12-05



MCP73831/2

NOTES:



APPENDIX A: REVISION HISTORY

Revision B (March 2006)

• Added MCP73832 through document.

Revision A (November 2005)

• Original Release of this Document.

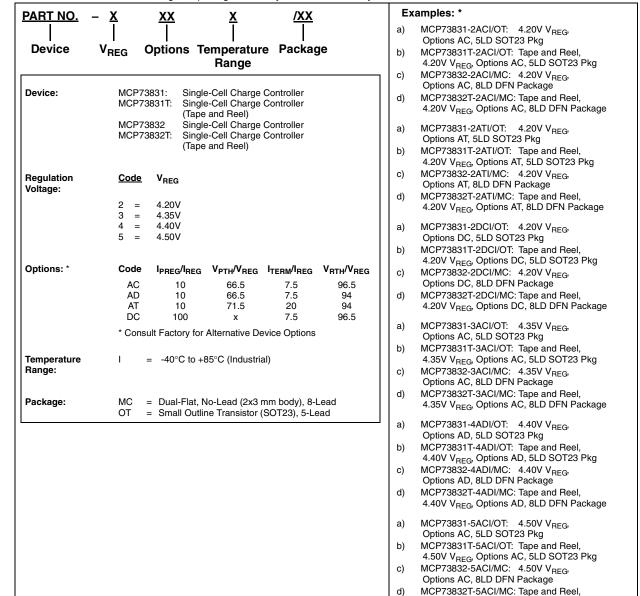
MCP73831/2

NOTES:



PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.



4.50V V_{REG}, Options AC, 8LD DFN Package

* Consult Factory for Alternate Device Options

MCP73831/2

NOTES:



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- Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
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